# Local Stabilization for Discrete-Time Systems with Distributed State Delay and Fast-Varying Input Delay under Actuator Saturations

Yonggang Chen and Zidong Wang

Abstract—This paper is concerned with the local stabilization problem for discrete-time systems with both distributed state delay and fastvarying input delay under actuator saturations. By introducing some terms concerning the distributedly delayed state and the current state, a novel polytopic model is first proposed to characterize the delayed saturation nonlinearity. Then, by incorporating a piecewise Lyapunov functional and some summation inequalities, a sufficient condition is established by means of linear matrix inequalities under which the closedloop system is locally exponentially stable. Moreover, the conditions for two special cases with single state delay and single input delay are proposed. Subsequently, certain optimization problems are formulated with aim to maximize the estimate of the region of attraction. Finally, two examples show the effectiveness and values of the obtained results.

Index Terms—Local stabilization, discrete-time systems, distributed state delay, fast-varying input delay, actuator saturations.

## I. INTRODUCTION

Time delays are often encountered in many real-world control systems and their existence is likely to lead to performance degradation or even instability of a control system. As such, the analysis and synthesis for time-delay systems have gained significant attention over the past three decades, see. e.g. [4], [12], [22], [24]. Overall, three types of time delays have been addressed in the literature, namely, discrete delays, distributed delays and neutral delays. In the context of linear matrix inequalities (LMIs), many advanced techniques have been developed for the convenience of controller/fiter design based on the utilization of some inequalities including Wirtinger-based inequalities [13], [14], free-matrix-based inequalities [25], [26], Bessel-Legendre inequalities [29] and reciprocally convex inequalities [15], [27]. However, it is noted that most available results have been concerned with a single discrete delay. For distributed delay systems, some recent results can be found in [3], [10], [11], [18].

Due to unavoidable physical constraints, actuator saturations are commonly encountered in practical feedback control systems, which is another source of performance degradation and system instability. For more than two decades, the stability/performance analysis and control design have been extensively studied for linear systems with saturating actuators [6], [7], [21], [30], [32]. If the open-loop system is not exponentially unstable, the semi-global/global stabilization can

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Z. Wang is with the College of Electrical Engineering and Automation, Shandong University of Science and Technology, Qingdao 266590, China, and also with the Department of Computer Science, Brunel University London, Uxbridge UB8 3PH, U.K. E-mail: Zidong.Wang@brunel.ac.uk be achieved via some typical approaches such as pole placement, parameterized Riccati equation and low-and-high gain design methods [7], [30]. In case the open-loop system is exponentially unstable, the local/regional analysis and synthesis can be carried out by using the polytopic models and the generalized sector condition [21], [32].

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For control systems subject to both time delays and actuator saturations, some pioneering results have appeared in the literature [1], [2], [5], [8], [9], [16], [17], [19], [20], [23], [28], [31]. For example, the local/regional stabilization problem has been investigated in [1], [5], [28] for continuous-time saturated systems with state delays. For discrete-time state-delayed systems with saturating actuators, some remarkable results have been proposed in [2], [16], [17], [19], [23]. In [9], the solution bounds have been obtained within the firstdelay-interval for input-delayed systems and the regional stabilization problem has been subsequently solved under actuator saturations. Nevertheless, it is worth pointing out that most existing results have been obtained for saturated control systems with a single state delay, input delay or output delay. Moreover, in [9], the time-varying input delay has been assumed to be *slowly-varying* and the technique used to handle the saturation nonlinearity is a bit conservative. Up to now, to the best of our knowledge, the local stabilization problem has not been sufficiently examined for saturated control systems with both state and input delays, not to mention the case that the distributed state delay and the fast-varying input delay are also involved.

Motivated by the above discussions, this paper is devoted to the investigation of the local stabilization problem for discrete-time systems with *both* distributed state delay *and* fast-varying input delay under actuator saturations. *The main contributions of the paper are summarized as follows.* 1) *The traditional constraint on the timevarying input delay is removed for saturated input-delay systems and a novel analysis approach is proposed.* 2) *By introducing the distributed delay terms and the current state term, a new polytopic model is proposed to characterize the delayed saturation nonlinearity.* 3) *The local stabilization condition and the optimization problem are established for saturated systems with both state and input delays.* 

**Notation.**  $P > 0 \ (\geq 0)$  denotes that P is a real, symmetric, and positive definite (positive semi-definite) matrix.  $\mathbb{R}^n$  is the *n*dimensional Euclidean space.  $\lambda_M(P)$  is the maximum eigenvalue of the matrix P.  $\|\cdot\|_2$  and  $\|\cdot\|_{\infty}$  are the 2-norm and  $\infty$ -norm of a vector, respectively.  $e_{m,k} \in \mathbb{R}^{1 \times m}$  is a row vector whose k-th element is 1 and others are zero.  $\otimes$  is the Kronecker product.  $K_{(l)}$  is the *l*-th row of the matrix K. I is the identity matrix with compatible dimension.

## II. PROBLEM FORMULATION

Consider the following discrete-time system with distributed state delay, fast-varying input delay and actuator saturations:

$$x(k+1) = Ax(k) + A_d \sum_{i=1}^{+\infty} \mu_i x(k-i) + Bsat(u(k-\tau_k)), \quad (1)$$

$$x(k) = \phi(k), \ k \in (-\infty, 0]$$
 (2)

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where  $x(k) \in \mathbb{R}^n$  denotes the system state;  $u(k) \in \mathbb{R}^m$  is the control input with u(k) = 0 for k < 0;  $\phi(k) \in \mathbb{R}^n$  is the initial condition;  $A, A_d$ , and B are known real constant matrices; the summation  $\sum_{i=1}^{+\infty} \mu_i x(k-i)$  represents the distributed state-delay term; and  $\tau_k$ denotes the time-varying input delay satisfying  $0 \le \tau_k \le \tau$  ( $\tau$  is a positive integer). Here,  $\operatorname{sat}(u) = [\operatorname{sat}(u_1) \operatorname{sat}(u_2) \cdots \operatorname{sat}(u_m)]^T$ is the standard saturation function with unity saturation level, where  $\operatorname{sat}(u_l) = \operatorname{sgn}(u_l)\min\{|u_l|, 1\}, l \in [1, m].$ 

The initial condition  $\phi(k)$  is assumed to belong to the set

$$\mathscr{X}_{\rho} \triangleq \left\{ \phi(k) : \max_{k \le 0} \|\phi(k)\|_2 \le \rho_1, \ \max_{k \le -1} \|\Delta\phi(k)\|_2 \le \rho_2 \right\}$$
(3)

where  $\Delta \phi(k) = \phi(k+1) - \phi(k)$ , and  $\rho_1$  and  $\rho_2$  are positive scalars. Of course, we can also assume that  $\phi(k)$  belongs to the set

$$\tilde{\mathscr{X}}_{\rho} \triangleq \{\phi(k) : \max_{k \le 0} \|\phi(k)\|_2 \le \rho\}.$$
(4)

Remark 1: For local stabilization of saturated systems, the exact characterization of the region of attraction is very difficult. In this paper, the estimate of region of attraction is characterized by the set  $\mathscr{X}_{\rho}$  or  $\mathscr{\tilde{X}}_{\rho}$ . For the set  $\mathscr{X}_{\rho}$ , the constrains on the 2-norms of  $\phi(k)$  and its variation  $\Delta \phi(k)$  are required. However, there is no requirement on  $\Delta \phi(k)$  for the set  $\mathscr{X}_{\rho}$ . If one sets  $\rho_2 = 2\rho_1 \triangleq \rho$  in  $\mathscr{X}_{\rho}$ , the set  $\mathscr{\tilde{X}}_{\rho}$  is recovered [2], [19]. When  $\phi(k)$  is slowly-varying with  $\rho_2 < 2\rho_1$ , the set  $\mathscr{X}_{\rho}$  is less conservative than  $\mathscr{\tilde{X}}_{\rho}$  in characterizing admissible initial conditions. Some other effective characterizations of admissible initial conditions have been presented in [16], [19].

In this paper, we employ the state feedback controller

$$u(k) = Kx(k), \ k \ge 0 \tag{5}$$

where  $K \in \mathbb{R}^{m \times n}$  is the controller gain matrix.

For the system (1), we make the following assumptions.

Assumption 1: For given scalars  $\mu_i > 0$  (i = 1, 2, ...), there exists a positive scalar  $0 < \lambda \le 1$  such that

$$\sum_{i=1}^{+\infty} \mu_i \lambda^{-i} < \sum_{j=1}^{+\infty} \sum_{i=1}^{j} \mu_j \lambda^{-i} < \sum_{l=1}^{+\infty} \sum_{j=1}^{l} \sum_{i=1}^{j} \mu_l \lambda^{-i} < +\infty.$$

Assumption 2: There exists a time sequence

$$= k_1 < \overline{k}_1 < k_2 < \overline{k}_2 < \dots < k_r < \overline{k}_r = k^* \le \tau$$

such that the following relationship holds:

$$k - \tau_k < 0, \ k \in \mathcal{T}_{\uparrow}; \ k - \tau_k \ge 0, \ k \in \mathcal{T}_{\downarrow} \cup [k^*, +\infty)$$

where

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$$\mathcal{T}_{\uparrow} \triangleq [k_1, \overline{k}_1) \cup [k_2, \overline{k}_2) \cup \dots \cup [k_r, \overline{k}_r), \mathcal{T}_{\downarrow} \triangleq [\overline{k}_1, k_2) \cup [\overline{k}_2, k_3) \cup \dots \cup [\overline{k}_{r-1}, k_r).$$

Remark 2: In Assumption 2, the time sequence does not have to be exactly known. Here, we only require that  $k^*$  is known (or unknown but upper-bounded by a known integer  $\tau^* \leq \tau$ ). In [9], it is assumed that there exists a *unique* integer  $k^*$  such that  $k - \tau_k < 0$  for  $k \in$  $[0, k^*)$  and  $k - \tau_k \geq 0$  for  $k \in [k^*, +\infty)$ . Such an assumption requires the input delay  $\tau_k$  to be slowly-varying with  $|\tau_{k+1} - \tau_k| \leq 1$ . From Assumption 2, it is seen that  $\tau_k$  can be fast-varying in this paper since the constraint  $|\tau_{k+1} - \tau_k| \leq 1$  is no longer necessary.

Under Assumption 2, there will be no control signal (i.e,  $u(k - \tau_k) = 0$ ) within the interval  $\mathcal{T}_{\uparrow}$  due to the time delay of the control input. In this case, the system (1) is equivalent to

$$x(k+1) = Ax(k) + A_d \sum_{i=1}^{+\infty} \mu_i x(k-i), \ k \in \mathcal{T}_{\uparrow}.$$
 (6)

To handle the actuator saturations within the interval  $\mathcal{T}_{\downarrow}$ , as in [6], we define the dead-zone nonlinearity  $\psi(u(k)) \triangleq u(k) - \operatorname{sat}(u(k))$ . Adding and subtracting  $u(k - \tau_k)$  in the right side of (1) yields

$$x(k+1) = Ax(k) + BKx(k-\tau_k) + A_d \sum_{i=1}^{+\infty} \mu_i x(k-i)$$
$$-B\psi(u(k-\tau_k)), \ k \in \mathcal{T}_{\downarrow}.$$
(7)

Moreover, the following (classical) sector condition holds [6], [21]:

$$\psi^{T}(u(k-\tau_{k}))H[\psi(u(k-\tau_{k})) - Kx(k-\tau_{k})] \le 0, \ k \in \mathcal{T}_{\downarrow}$$
(8)

where  $H \in \mathbb{R}^{m \times m}$  is any positive diagonal matrix.

Next, we will introduce two important lemmas.

Lemma 1: [32] Let  $v \in \mathbb{R}^{\overrightarrow{m}}$  be such that  $||v||_{\infty} \leq 1$  where  $\overleftarrow{m} = m2^{m-1}$ . Let the elements in  $\mathbb{D}_m$  be labeled as  $D_i$   $(i \in [1, 2^m])$  where  $\mathbb{D}_m$  is a set of  $m \times m$  diagonal matrices with diagonal elements being either 1 or 0, and the function  $f_m$  be defined as  $f_m(0) = 0$  and

$$f_m(i) = \begin{cases} f_m(i-1) + 1, & D_i + D_j \neq I_m, & \forall j \in [1,i] \\ f_m(j), & D_i + D_j = I_m, & \exists j \in [1,i] \end{cases}.$$

Then, for any  $u \in \mathbb{R}^m$ , there holds  $\operatorname{sat}(u) \in \operatorname{co}\{D_i u + \mathcal{D}_i^- v : i \in [1, 2^m]\}$ , where "co" denotes the convex hull and  $\mathcal{D}_i^- \in \mathbb{R}^{m \times \overrightarrow{m}}$  is defined as  $\mathcal{D}_i^- = e_{2^{m-1}, f_m(i)} \otimes D_i^-$  with  $D_i^- = I - D_i$ .

Lemma 2: [2], [18] Let  $0 < Z \in \mathbb{R}^{n \times n}$ ,  $x_i \in \mathbb{R}^n$  and the scalars  $\mu_i \ge 0$ ,  $\lambda_j \ge 0$   $(i, j = 1, 2, \cdots)$  be given. Then, we have

1) 
$$\left(\sum_{i=1}^{+\infty} \mu_{i} x_{i}\right)^{T} Z\left(\sum_{i=1}^{+\infty} \mu_{i} x_{i}\right)$$
$$\leq \left(\sum_{i=1}^{+\infty} \mu_{i} \lambda_{i}^{-1}\right) \left(\sum_{i=1}^{+\infty} \mu_{i} \lambda_{i} x_{i}^{T} Z x_{i}\right),$$
$$2) \quad \left(\sum_{j=1}^{+\infty} \sum_{i=1}^{j} \mu_{j} x_{i}\right)^{T} Z\left(\sum_{j=1}^{+\infty} \sum_{i=1}^{j} \mu_{j} x_{i}\right)$$
$$\leq \left(\sum_{j=1}^{+\infty} \sum_{i=1}^{j} \mu_{j} \lambda_{i}^{-1}\right) \left(\sum_{j=1}^{+\infty} \sum_{i=1}^{j} \mu_{j} \lambda_{i} x_{i}^{T} Z x_{i}\right).$$

Now, let U, V and W be  $\stackrel{\leftrightarrow}{m} \times n$  matrices and denote that

$$v(k) \triangleq Ux(k) + V \sum_{i=1}^{+\infty} \mu_i x(k-i) + W \sum_{j=1}^{+\infty} \mu_j \sum_{i=k-j}^{k-1} x(i).$$
(9)

Assume that the following constraint condition holds:

$$\|v(k)\|_{\infty} \le 1, \ k \in [k^*, +\infty).$$
 (10)

Then, from Lemma 1,  $sat(u(k - \tau_k))$  can be represented as

$$\operatorname{sat}(u(k-\tau_k)) = \sum_{s=1}^{2^m} \varpi_s^k \left[ D_s u(k-\tau_k) + \mathcal{D}_s^- v(k) \right]$$
(11)

where  $k \in [k^*, +\infty)$ ,  $\varpi_s^k \ge 0$  ( $s \in [1, 2^m]$ ) and  $\sum_{s=1}^{2^m} \varpi_s^k = 1$ . We are now ready to present the closed-loop system as follows:

$$x(k+1) = \sum_{s=1}^{2^{m}} \varpi_{s}^{k} \left\{ (A + B\mathcal{D}_{s}^{-}U)x(k) + BD_{s}K \times x(k-\tau_{k}) + (A_{d} + B\mathcal{D}_{s}^{-}V) \sum_{i=1}^{+\infty} \mu_{i}x(k-i) + B\mathcal{D}_{s}^{-}W \sum_{j=1}^{+\infty} \mu_{j} \sum_{i=k-j}^{k-1} x(i) \right\}, \ k \in [k^{*}, +\infty).$$
(12)

*Remark 3:* To deal with the delayed saturation nonlinearity in a less conservative framework, the distributed-delay-dependent terms

 $\mathcal{D}_s^- V \sum_{i=1}^h \mu_i x(k-i)$  and  $\mathcal{D}_s^- W \sum_{j=1}^h \mu_j \sum_{i=k-j}^{k-1} x(i)$  are additionally introduced in the polytopic model (11). Furthermore, it is worth mentioning that, different from the sector condition used in [4] (pp. 239), our proposed polytopic model (11) utilizes the *current state* x(k) of the system (1), thereby facilitating the reduction of the possible conservatism. In addition, it is worth mentioning that the closed-loop system has different representations within the intervals  $\mathcal{T}_{\downarrow}$  and  $[k^*, +\infty)$ . Using the model (7), the global analysis can be performed within  $\mathcal{T}_{\downarrow}$  and using (12), the local analysis can be done in a less conservative manner for the multiple-input case [32].

This paper aims to design the controller (5) such that the closedloop system (12) is locally exponentially stable with an estimate of the region of attraction that is made as large as possible.

# III. MAIN RESULTS

For the purpose of exponential stability analysis, we first propose the following piecewise augmented Lyapunov functional:

$$V(k) = \begin{cases} V_1(k), \ k \in [k^*, +\infty), \\ V_2(k), \ k \in [0, k^*) \end{cases}$$
(13)

where

$$V_{\alpha}(k) = \eta^{T}(k)P_{\alpha}\eta(k) + \sum_{i=k-\tau}^{k-1} \lambda_{\alpha}^{k-i-1}x^{T}(i)Q_{\alpha}x(i) + \sum_{j=1}^{+\infty} \mu_{j} \sum_{i=k-j}^{k-1} \lambda_{\alpha}^{k-i-1}x^{T}(i)S_{\alpha 1}x(i) + \sum_{l=1}^{+\infty} \mu_{l} \sum_{j=1}^{l} \sum_{i=k-j}^{k-1} \lambda_{\alpha}^{k-i-1}x^{T}(i)S_{\alpha 2}x(i) + \tau \sum_{j=-\tau}^{-1} \sum_{i=k+j}^{k-1} \lambda_{\alpha}^{k-i-1}y^{T}(i)(R_{\alpha 1} + R_{\alpha 2})y(i) + \sum_{l=1}^{+\infty} \mu_{l} \sum_{j=1}^{l} \sum_{i=k-j}^{k-1} \lambda_{\alpha}^{k-i-1}y^{T}(i)Z_{\alpha}y(i), \ \alpha = 1, 2$$

with  $P_{\alpha} > 0, \ Q_{\alpha} > 0, \ S_{\alpha 1} > 0, \ S_{\alpha 2} > 0, \ R_{\alpha 1} > 0, \ R_{\alpha 2} > 0, \ Z_{\alpha} > 0, \ 0 < \lambda_1 \le 1, \ \lambda_2 > 1, \ y(k) = x(k+1) - x(k), \text{ and } \eta(k) = \left[x^T(k) \sum_{i=k-\tau}^{k-1} x^T(i) \sum_{j=1}^{+\infty} \mu_j \sum_{i=k-j}^{k-1} x^T(i)\right]^T.$ For convenience of subsequent presentation, we define

$$\begin{split} \Gamma_{3} &\triangleq \begin{bmatrix} I & 0 & 0 & 0 & 0 & -I & \overline{\tau} \\ 0 & 0 & -I & \overline{\tau} I & 0 & 0 & 0 & 0 \\ \kappa I & 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ \kappa I & 0 & 0 & \overline{\tau} I & 0 & 0 & 0 & 0 \\ -I & 0 & 0 & \overline{\tau} I & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & I & 0 \\ \kappa I & 0 & 0 & -I & I & 0 \end{bmatrix}, \\ \Gamma_{5} &\triangleq \begin{bmatrix} I & 0 & 0 & 0 & 0 & 0 \\ 0 & -I & \overline{\tau} I & 0 & 0 & 0 \\ \kappa I & 0 & 0 & -I & I & 0 \end{bmatrix}, \\ \Phi_{1} &\triangleq \begin{bmatrix} I & -I & 0 & 0 & 0 & 0 & 0 \\ -I & 0 & \overline{\tau} I & 0 & 0 & 0 & 0 \\ \kappa I & 0 & 0 & 0 & I & 0 \end{bmatrix}, \\ \Phi_{2} &\triangleq \begin{bmatrix} \kappa I & 0_{n \times 4n} & -I & 0_{n \times 2n} \\ 1 & I & 0 & -2I & 0 & 0_{n \times 3n} \\ 0 & I & I & 0 & -2I & 0_{n \times 3n} \\ 0 & I & -I & 0 & 0_{n \times (4n+m)} \end{bmatrix} \end{bmatrix}, \\ \Phi_{4} &\triangleq \begin{bmatrix} I & 0 & -I & 0 & 0_{n \times (4n+m)} \\ I & 0 & I & -2I & 0_{n \times (3n+m)} \\ I & 0 & I & -2I & 0_{n \times (3n+m)} \end{bmatrix}, \\ \Phi_{5} &\triangleq \begin{bmatrix} \kappa I & 0_{n \times 2n} & -I & 0_{n \times (2n+m)} \\ I & 0 & I & -2I & 0_{n \times (3n+m)} \end{bmatrix}, \\ \Phi_{6} &\triangleq \begin{bmatrix} I & -I & 0 & 0_{n \times 2n} \\ I & I & -2I & 0_{n \times 3n} \end{bmatrix}, R_{\alpha} \triangleq R_{\alpha 1} + R_{\alpha 2}, \\ \Phi_{7} &\triangleq \begin{bmatrix} \kappa I & 0_{n \times 2n} & -I & 0_{n \times 2n} \\ I & I & 0 & -2I & 0_{n \times 2n} \end{bmatrix}, R_{\alpha} \triangleq \bar{R}_{\alpha 1} + \bar{R}_{\alpha 2}, \\ \Phi_{7} &\triangleq \begin{bmatrix} \kappa I & 0_{n \times 2n} & -I & 0_{n \times 2n} \\ I & I & -2I & 0_{n \times 3n} \end{bmatrix}, R_{\alpha} \triangleq \bar{R}_{\alpha 1} + \bar{R}_{\alpha 2}, \\ \Phi_{7} &\triangleq \begin{bmatrix} \kappa I & 0_{n \times 2n} & -I & 0_{n \times 2n} \\ I & I & -2I & 0_{n \times 3n} \end{bmatrix}, R_{\alpha} \triangleq \bar{R}_{\alpha 1} + \bar{R}_{\alpha 2}, \\ \Phi_{7} &\triangleq \begin{bmatrix} \kappa I & 0_{n \times 2n} & -I & 0_{n \times 2n} \\ I & I & 0 & 1 & 2I & 0_{n \times 2n} \end{bmatrix}, \bar{R}_{\alpha} \triangleq \bar{R}_{\alpha 1} + \bar{R}_{\alpha 2}, \\ \Phi_{7} &\triangleq \begin{bmatrix} \kappa I & 0_{n \times 2n} & -I & 0_{n \times 2n} \\ I & I & 0 & 1 & 2I & 0_{n \times 2n} \end{bmatrix}, R_{\alpha} \triangleq \bar{R}_{\alpha 1} + \bar{R}_{\alpha 2}, \\ \Phi_{7} &\triangleq \begin{bmatrix} \kappa I & 0_{n \times 2n} & -I & 0_{n \times 2n} \\ I & I & 0 & 0 & 0 & 0 & 0 \end{bmatrix}, \\ \tilde{R}_{\alpha} &\triangleq \bar{R}_{\alpha} = \bar{R}_{\alpha} =$$

Theorem 1: Let the scalars  $0 < \lambda_1 \leq 1, \lambda_2 > 1, \nu > 0$ , the integer  $k^* \geq 1$  and the matrix K be given. Assume that there exist matrices  $0 < P_{\alpha} \in \mathbb{R}^{3n \times 3n}, 0 < Q_{\alpha} \in \mathbb{R}^{n \times n}, 0 < S_{\alpha j} \in \mathbb{R}^{n \times n}, 0 < R_{\alpha j} \in \mathbb{R}^{n \times n}, 0 < Z_{\alpha} \in \mathbb{R}^{n \times n}, T_{ij} \in \mathbb{R}^{n \times n}$   $(\alpha, j = 1, 2, 3), M_1 \in \mathbb{R}^{2n \times 2n}, M_2 \in \mathbb{R}^{n \times n}, U \in \mathbb{R}^{\overrightarrow{m} \times n}, V \in \mathbb{R}^{\overrightarrow{m} \times n}, W \in \mathbb{R}^{\overrightarrow{m} \times n}, \text{ and the diagonal matrix } 0 < H \in \mathbb{R}^{m \times m}$  such that, for  $\tau_k = 0, \tau, \forall s \in [1, 2^m], \forall l \in [1, \overrightarrow{m}]$ , the matrix inequalities

$$\Lambda_{1} \triangleq \begin{bmatrix} \check{R}_{1} & M_{1} \\ M_{1}^{T} & \check{R}_{1} \end{bmatrix} > 0, \ \Lambda_{2} \triangleq \begin{bmatrix} R_{21} & M_{2} \\ M_{2}^{T} & R_{21} \end{bmatrix} > 0,$$
(14)  
$$\Xi_{1}(\tau_{k}, s) \triangleq \Gamma_{1}^{T} P_{1}\Gamma_{1} - \lambda_{1}\Gamma_{2}^{T} P_{1}\Gamma_{2} - \lambda_{1}^{\tau}\Phi_{1}^{T}\Lambda_{1}\Phi_{1}$$

$$-\Phi_2^T(Z_1/\tilde{\sigma}_1)\Phi_2 + \operatorname{Sym}(T_1\Sigma_1) + \Psi_1 < 0, \tag{15}$$

$$\Xi_{2} \equiv \Gamma_{3}^{T} P_{2} \Gamma_{3} - \lambda_{2} (\Gamma_{4}^{T} P_{2} \Gamma_{4} + \Phi_{3}^{T} \Lambda_{2} \Phi_{3} + \Phi_{4}^{T} R_{22}^{2} \Phi_{4}) - \Phi_{5}^{T} (Z_{2} / \tilde{\sigma}_{2}) \Phi_{5} + \operatorname{Sym}(T_{2} \Sigma_{2} + T_{4} \Sigma_{4}) + \Psi_{2} < 0, \quad (16) \Xi_{3} \triangleq \Gamma_{5}^{T} P_{2} \Gamma_{5} - \lambda_{2} \Gamma_{6}^{T} P_{2} \Gamma_{6} - \lambda_{2} \Phi_{6}^{T} \tilde{R}_{2}^{T} \Phi_{6}$$

$$\begin{aligned}
\mathcal{L}_{3} &= \Gamma_{5}^{T} P_{2} \Gamma_{5} - \lambda_{2} \Gamma_{6}^{T} P_{2} \Gamma_{6} - \lambda_{2} \Phi_{6}^{T} R_{2}^{\prime} \Phi_{6} \\
&- \Phi_{7}^{T} (Z_{2} / \tilde{\sigma}_{2}) \Phi_{7} + \operatorname{Sym}(T_{3} \Sigma_{3}) + \Psi_{3} < 0,
\end{aligned} \tag{17}$$

$$\begin{cases} P_1 \le \nu P_2, \ Q_1 \le \nu Q_2, \ S_{1j} \le \nu S_{2j}, \end{cases}$$
(18)

$$\left( \begin{array}{c} R_{1j} \le \nu R_{2j} \ (j = 1, 2), \ Z_1 \le \nu Z_2, \\ \Xi_4(l) \triangleq \begin{bmatrix} 1/(\nu \lambda_2^{k^*}) & N_{(l)} \\ N_{(l)}^T & \text{diag}\{P_1, 0\} + \Psi_4/\lambda_1 \end{bmatrix} \ge 0,$$
 (19)

e satisfied, where 
$$N_{(l)} = [U_{(l)} \ 0 \ W_{(l)} \ V_{(l)}]$$
 and  
 $\Psi_1 = \text{diag}\{Q_1 + \kappa S_{11} + \sigma S_{12}, 0, -\lambda_1^T Q_1, 0, 0, -S_{11}/\tilde{\kappa}_1, -S_{12}/\tilde{\sigma}_1, \tau^2 R_1 + \sigma Z_1\},$   
 $\Psi_2 = \text{diag}\{Q_2 + \kappa S_{21} + \sigma S_{22}, 0, -\lambda_2^T Q_2, 0, -S_{21}/\tilde{\kappa}_2, -S_{22}/\tilde{\sigma}_2, \tau^2 R_2 + \sigma Z_2, 0\},$   
 $\Psi_3 = \text{diag}\{Q_2 + \kappa S_{21} + \sigma S_{22}, -\lambda_2^T Q_2, 0, -S_{21}/\tilde{\kappa}_2, -S_{22}/\tilde{\sigma}_2, \tau^2 R_2 + \sigma Z_2\},$   
 $\Psi_4 = \text{diag}\{0, (\lambda_1^T/\tau)Q_1, S_{12}/\tilde{\sigma}_1, S_{11}/\tilde{\kappa}_1\} + (2\lambda_1^T/\tilde{\tau})\Phi_8^T R_1 \Phi_8 + (1/\tilde{\sigma}_1)\Phi_9 Z_1 \Phi_9,$   
 $T_1 = [T_{11}^T \ 0_{n \times 6n} \ T_{12}^T]^T, T_3 = [T_{31}^T \ 0_{n \times 4n} \ T_{32}^T]^T,$   
 $T_2 = [T_{21}^T \ 0_{n \times 5n} \ T_{22}^T \ 0_{n \times m}]^T, T_4 = [0_{m \times 7n} \ H^T]^T,$   
 $\Sigma_1 = [A + BD_s^- V \ BD_s^- W \ -I],$   
 $\Sigma_2 = [A - I \ BK \ 0_{n \times 2n} \ A_d \ 0_{n \times n} \ -I], \beta_1 = \text{diag}\{R_1, 3R_1\},$   
 $\tilde{K}_2^T = \text{diag}\{R_2, 3\varphi_\tau R_2\}, \ \tilde{K}_{22}^T \equiv \text{diag}\{R_{22}, 3\varphi_\tau R_{22}\},$   
 $(\varphi_\tau \triangleq (\tau + 1)/(\tau - 1) \ (\tau > 1), \ \varphi_1 \triangleq 1).$ 

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Then, for any initial condition  $\phi(k) \in \mathscr{X}_{\rho}$  satisfying  $V_2(0) \leq 1$ , the closed-loop system (12) is exponentially stable.

*Proof:* Letting  $\triangle V_{\alpha}(k) \triangleq V_{\alpha}(k+1) - \lambda_{\alpha}V_{\alpha}(k)$  ( $\alpha = 1, 2$ ), by calculations and using Lemma 2, it follows that

$$\Delta V_{\alpha}(k) \leq \eta^{T}(k+1)P_{\alpha}\eta(k+1) - \lambda_{\alpha}\eta^{T}(k)P_{\alpha}\eta(k) + x^{T}(k)(Q_{\alpha} + \kappa S_{\alpha 1} + \sigma S_{\alpha 2})x(k) + y^{T}(k) \times (\tau^{2}R_{\alpha} + \sigma Z_{\alpha})y(k) - \lambda_{\alpha}^{\tau}x^{T}(k-\tau)Q_{\alpha}x(k-\tau) - \zeta_{1}^{T}(k)(S_{\alpha 1}/\tilde{\kappa}_{\alpha})\zeta_{1}(k) - \zeta_{2}^{T}(k)(S_{\alpha 2}/\tilde{\sigma}_{\alpha})\zeta_{2}(k) - \tau\tilde{\lambda}_{\alpha}\sum_{i=k-\tau}^{k-1} y^{T}(i)R_{\alpha}y(i) - \zeta_{3}^{T}(k)(Z_{\alpha}/\tilde{\sigma}_{\alpha})\zeta_{3}^{T}(k)$$
(20)

where  $\tilde{\lambda}_1 = \lambda_1^{\tau}$ ,  $\tilde{\lambda}_2 = \lambda_2$ ,  $\zeta_1(k) = \sum_{i=1}^{+\infty} \mu_i x(k-i)$ ,  $\zeta_2(k) = \sum_{j=1}^{+\infty} \sum_{i=1}^{j} \mu_j x(k-i)$ , and  $\zeta_3(k) = \sum_{j=1}^{+\infty} \sum_{i=1}^{j} \mu_j y(k-i)$ . Using the discrete Wirtinger-based inequality [2], [14] and the

Using the discrete Wirtinger-based inequality [2], [14] and the reciprocally convex combination inequality [14], and noting the fact  $\sum_{i=k-\tau}^{k-1} (\cdot) = \sum_{i=k-\tau_k}^{k-1} (\cdot) + \sum_{i=k-\tau}^{k-\tau_k-1} (\cdot)$ , we have

$$\tau \sum_{i=k-\tau}^{k-1} y^T(i) R_1 y(i) \ge \xi_1^T(k) \Phi_1^T \Lambda_1 \Phi_1 \xi_1(k)$$
(21)

subject to the constraint  $\Lambda_1 > 0$  (the first inequality in (14)). Similarly, using the Jensen inequality [2] and the reciprocally convex combination inequality [14], it follows that

$$\tau \sum_{i=k-\tau}^{k-1} y^T(i) R_{21} y(i) \ge \xi_2^T(k) \Phi_3^T \Lambda_2 \Phi_3 \xi_2(k)$$
(22)

where the matrix  $\Lambda_2$  given in (14) satisfies  $\Lambda_2 > 0$ . In particular, using Wirtinger-based inequality directly, one obtains

$$\tau \sum_{i=k-\tau}^{k-1} y^{T}(i) R_{22} y(i) \ge \xi_{2}^{T}(k) \Phi_{4}^{T} \check{R}_{22}^{\tau} \Phi_{4} \xi_{2}(k), \qquad (23)$$

$$\tau \sum_{i=k-\tau}^{k-1} y^{T}(i) R_{2} y(i) \ge \xi_{3}^{T}(k) \Phi_{6}^{T} \check{R}_{2}^{\tau} \Phi_{6} \xi_{3}(k).$$
(24)

Noting the matrices  $T_i$  and  $\Sigma_i$  (i = 1, 2, 3) that are denoted in the statement of the theorem, we obtain from (12), (7) and (6) that

$$2\xi_1^T(k)T_1\Sigma_1\xi_1(k) = 0, \ k \in [k^*, +\infty),$$
(25)

$$2\xi_2^T(k)T_2\Sigma_2\xi_2(k) = 0, \ k \in \mathcal{T}_{\downarrow},\tag{26}$$

$$2\xi_3^T(k)T_3\Sigma_3\xi_3(k) = 0, \ k \in \mathcal{T}_{\uparrow}.$$
(27)

In addition, it is seen from the sector condition (8) that

$$-2\xi_2^T(k)T_4\Sigma_4\xi_2(k) \ge 0, \ k \in \mathcal{T}_{\downarrow}.$$
(28)

Adding the left-hand side of (25) to  $\Delta V_1(k)$ , and using (21) and  $\sum_{j=1}^{+\infty} \sum_{i=1}^{j} \mu_j y(k-i) = \kappa x(k) - \sum_{i=1}^{+\infty} \mu_j x(k-i)$ , we obtain

$$\Delta V_1(k) \le \sum_{s=1}^{2^m} \varpi_s^k \xi_1^T(k) \Xi_1(\tau_k, s) \xi_1(k), \ k \in [k^*, +\infty)$$
(29)

where  $\Xi_1(\tau_k, s)$  is defined in (15). Similarly, adding the left-hand sides of (26) and (28) to  $\Delta V_2(k)$  and using (22)-(23), we have

$$\Delta V_2(k) \le \xi_2^T(k) \Xi_2 \xi_2(k), \ k \in \mathcal{T}_{\downarrow}$$
(30)

where  $\Xi_2$  is given in the matrix inequality (16). Also, adding the left side of (27) to  $\Delta V_2(k)$  and applying the inequality (24) yields

$$\Delta V_2(k) \le \xi_3^T(k) \Xi_3 \xi_3(k), \ k \in \mathcal{T}_{\uparrow}$$
(31)

where the matrix  $\Xi_3$  is denoted in the inequality (17).

For  $\tau_k = 0, \tau$  and  $\forall s \in [1, 2^m]$ , if the matrix inequalities (14)-(17) are satisfied, then we can obtain from (29)-(31) that

$$V_1(k+1) \le \lambda_1 V_1(k), \ k \in [k^*, +\infty),$$
 (32)

$$V_2(k+1) \le \lambda_2 V_2(k), \ k \in [0, k^*).$$
 (33)

Moreover, it is seen from the matrix inequalities in (18) that

$$V_1(k) \le \nu V_2(k), \ k \ge 0.$$
 (34)

Using the inequalities (32)-(34), it follows that

$$V_1(k) \le \lambda_1^{k-k^*} [\nu \lambda_2^{k^*} V_2(0)], \ k \in [k^*, +\infty).$$
 (35)

On the other hand, noting the following facts:

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$$\sum_{j=1}^{+\infty} \mu_j \sum_{i=k-j}^{k-1} f(i) \ge \sum_{j=1}^{+\infty} \mu_j f(k-j),$$
$$\sum_{l=1}^{+\infty} \mu_l \sum_{j=1}^{l} \sum_{i=k-j}^{k-1} f(i) \ge \sum_{l=1}^{+\infty} \mu_l \sum_{j=1}^{l} f(k-j)$$

where f(i) is a positive real function, and using Lemma 2 and the Jensen inequality [2], one has the following inequality:

$$V_1(k) \ge \tilde{\eta}^T(k) [\operatorname{diag}\{P_1, 0\} + \Psi_4/\lambda_1] \tilde{\eta}(k), \ k \in [k^*, +\infty)$$
 (36)

where  $\Psi_4$  is given in (19) and  $\tilde{\eta}(k) \triangleq \left[\eta^T(k) \sum_{i=1}^{+\infty} \mu_i x^T(k-i)\right]^T$ . Applying Schur complement to (19), it is clear that

diag{
$$P_1, 0$$
} +  $\Psi_4/\lambda_1 \ge \nu \lambda_2^{k^*} N_{(l)}^T N_{(l)}, \ l \in [1, \stackrel{\leftrightarrow}{m}].$  (37)

Then, one obtains from (35)-(37) that

$$|v_{l}(k)|^{2} = \tilde{\eta}^{T}(k) N_{(l)}^{T} N_{(l)} \tilde{\eta}(k) \leq (1/(\nu \lambda_{2}^{k^{*}})) V_{1}(k),$$
  
$$\leq V_{2}(0), \ l \in [1, \dot{\tilde{m}}], \ k \in [k^{*}, +\infty).$$
(38)

For any initial condition  $\phi(k) \in \mathscr{X}_{\rho}$  satisfying  $V_2(0) \leq 1$ , it is clear from (38) that the constraint condition (10) is ensured. Moreover, one can conclude from (35) that the closed-loop system (12) is locally exponentially stable and this completes the proof.

*Remark 4:* In the proof of Theorem 1, the global analysis is performed within the interval  $\mathcal{T}_{\downarrow}$  by using the classical sector condition (8) since  $\mathcal{T}_{\downarrow}$  is generally small and not exactly known. Considering that the open-loop of the system (1) might be unstable, the functional  $V_2(k)$  along the closed-loop system (7) is required to be increasing.

Of course, if the sequence  $k_1, \overline{k}_1, k_2, \overline{k}_2, \dots, k_r, \overline{k}_r$  in Assumption 2 is known, one can perform the less conservative local analysis.

Next, we will discuss the controller design in terms of LMIs. Theorem 2: Let the scalars  $0 < \lambda_1 \leq 1, \lambda_2 > 1, \nu > 0, \delta_i \neq 0$  (i = 1, 2, 3) and the integer  $k^* \geq 1$  be given. Assume that there exist matrices  $0 < \bar{P}_{\alpha} \in \mathbb{R}^{3n \times 3n}, 0 < \bar{Q}_{\alpha} \in \mathbb{R}^{n \times n}, 0 < \bar{S}_{\alpha j} \in \mathbb{R}^{n \times n}, 0 < \bar{R}_{\alpha j} \in \mathbb{R}^{n \times n}, 0 < \bar{Z}_{\alpha} \in \mathbb{R}^{n \times n}, (\alpha, j = 1, 2), \bar{M}_1 \in \mathbb{R}^{2n \times 2n}, \bar{M}_2 \in \mathbb{R}^{n \times n}, X \in \mathbb{R}^{n \times n}, Y \in \mathbb{R}^{m \times n}, \bar{U} \in \mathbb{R}^{\bar{m} \times n}, \bar{V} \in \mathbb{R}^{\bar{m} \times n}, \bar{W} \in \mathbb{R}^{\bar{m} \times n}, \text{ and the diagonal matrix } 0 < \bar{H} \in \mathbb{R}^{m \times m}$  such that for  $\tau_k = 0, \tau, \forall s \in [1, 2^m], \forall l \in [1, \bar{m}]$ , the following LMIs hold:

$$\bar{\Lambda}_{1} \triangleq \begin{bmatrix} \hat{R}_{1} & \bar{M}_{1} \\ \bar{M}_{1}^{T} & \hat{R}_{1} \end{bmatrix} > 0, \ \bar{\Lambda}_{2} \triangleq \begin{bmatrix} \bar{R}_{21} & \bar{M}_{2} \\ \bar{M}_{2}^{T} & \bar{R}_{21} \end{bmatrix} > 0,$$
(39)

$$\bar{\Xi}_1(\tau_k, s) \triangleq \Gamma_1^T \bar{P}_1 \Gamma_1 - \lambda_1 \Gamma_2^T \bar{P}_1 \Gamma_2 - \lambda_1^\tau \Phi_1^T \bar{\Lambda}_1 \Phi_1 - \Phi_2^T (\bar{Z}_1 / \tilde{\sigma}_1) \Phi_2 + \operatorname{Sym}(\bar{T}_1 \bar{\Sigma}_1) + \bar{\Psi}_1 < 0,$$

$$(40)$$

$$\bar{\Xi}_{2} \triangleq \Gamma_{3}^{T} \bar{P}_{2} \Gamma_{3} - \lambda_{2} (\Gamma_{4}^{T} \bar{P}_{2} \Gamma_{4} + \Phi_{3}^{T} \bar{\Lambda}_{2} \Phi_{3} + \Phi_{4}^{T} \hat{R}_{22}^{\tau} \Phi_{4})$$

$$-\Phi_{5}^{T}(Z_{2}/\tilde{\sigma}_{2})\Phi_{5} + \operatorname{Sym}(T_{2}\Sigma_{2} + T_{4}\Sigma_{4}) + \Psi_{2} < 0, \quad (41)$$
$$\bar{\Xi}_{3} \triangleq \Gamma_{5}^{T}\bar{P}_{2}\Gamma_{5} - \lambda_{2}\Gamma_{6}^{T}\bar{P}_{2}\Gamma_{6} - \lambda_{2}\Phi_{6}^{T}\hat{R}_{2}^{T}\Phi_{6}$$

$$-\Phi_7^T(\bar{Z}_2/\tilde{\sigma}_2)\Phi_7 + \text{Sym}(\bar{T}_3\bar{\Sigma}_3) + \bar{\Psi}_3 < 0, \tag{42}$$

$$\begin{cases} \bar{P}_1 \le \nu \bar{P}_2, \ \bar{Q}_1 \le \nu \bar{Q}_2, \ \bar{S}_{1j} \le \nu \bar{S}_{2j}, \\ \bar{R}_{1j} \le \nu \bar{R}_{2j} \ (j=1,2), \ \bar{Z}_1 \le \nu \bar{Z}_2, \end{cases}$$
(43)

$$\bar{\Xi}_4(l) \triangleq \begin{bmatrix} 1/(\nu \lambda_2^{k^*}) & \bar{N}_{(l)} \\ \bar{N}_{(l)}^T & \text{diag}\{\bar{P}_1, 0\} + \bar{\Psi}_4/\lambda_1 \end{bmatrix} \ge 0$$
(44)

where  $\bar{N}_{(l)} = [\bar{U}_{(l)} \ 0 \ \bar{W}_{(l)} \ \bar{V}_{(l)}]$  and

$$\begin{split} \bar{\Psi}_{1} = & \operatorname{diag}\{\bar{Q}_{1} + \kappa \bar{S}_{11} + \sigma \bar{S}_{12}, 0, -\lambda_{1}^{\tau} \bar{Q}_{1}, 0, \\ 0, -\bar{S}_{11}/\tilde{\kappa}_{1}, -\bar{S}_{12}/\tilde{\sigma}_{1}, \tau^{2} \bar{R}_{1} + \sigma \bar{Z}_{1}\}, \\ \bar{\Psi}_{2} = & \operatorname{diag}\{\bar{Q}_{2} + \kappa \bar{S}_{21} + \sigma \bar{S}_{22}, 0, -\lambda_{2}^{\tau} \bar{Q}_{2}, 0, \\ -\bar{S}_{21}/\tilde{\kappa}_{2}, -\bar{S}_{22}/\tilde{\sigma}_{2}, \tau^{2} \bar{R}_{2} + \sigma \bar{Z}_{2}, 0\}, \\ \bar{\Psi}_{3} = & \operatorname{diag}\{\bar{Q}_{2} + \kappa \bar{S}_{21} + \sigma \bar{S}_{22}, -\lambda_{2}^{\tau} \bar{Q}_{2}, 0, \\ -\bar{S}_{21}/\tilde{\kappa}_{2}, -\bar{S}_{22}/\tilde{\sigma}_{2}, \tau^{2} \bar{R}_{2} + \sigma \bar{Z}_{2}\}, \\ \bar{\Psi}_{4} = & \operatorname{diag}\{0, (\lambda_{1}^{\tau}/\tau) \bar{Q}_{1}, \bar{S}_{12}/\tilde{\sigma}_{1}, \bar{S}_{11}/\tilde{\kappa}_{1}\} \\ + (2\lambda_{1}^{\tau}/\tilde{\tau}) \Phi_{8}^{T} \bar{R}_{1} \Phi_{8} + (1/\tilde{\sigma}_{1}) \Phi_{9} \bar{Z}_{1} \Phi_{9}, \\ \bar{T}_{1} = & \begin{bmatrix} I & 0_{n \times 6n} & \delta_{1}I \end{bmatrix}^{T}, \ \bar{T}_{3} = & \begin{bmatrix} I & 0_{n \times 4n} & \delta_{3}I \end{bmatrix}^{T}, \\ \bar{T}_{2} = & \begin{bmatrix} I & 0_{n \times 5n} & \delta_{2}I & 0_{n \times m} \end{bmatrix}^{T}, \ \bar{T}_{4} = & \begin{bmatrix} 0_{m \times 7n} & I \end{bmatrix}^{T}, \\ \bar{\Sigma}_{1} = & \begin{bmatrix} (A - I)X^{T} + BD_{s}^{-} \bar{U} & BD_{s}Y & 0_{n \times 3n} \\ A_{d}X^{T} + BD_{s}^{-} \bar{V} & BD_{s}^{-} \bar{W} & -X^{T} \end{bmatrix}, \\ \bar{\Sigma}_{2} = & \begin{bmatrix} (A - I)X^{T} & 0_{n \times 2n} & A_{d}X^{T} & 0_{n \times n} & -X^{T} \end{bmatrix}, \\ \bar{\Sigma}_{3} = & \begin{bmatrix} (A - I)X^{T} & 0_{n \times 2n} & A_{d}X^{T} & 0_{n \times n} & -X^{T} \end{bmatrix}, \\ \bar{\Sigma}_{4} = & \begin{bmatrix} 0_{m \times n} & Y & 0_{m \times 5n} & -\bar{H} \end{bmatrix}, \ \hat{R}_{1} \triangleq & \operatorname{diag}\{\bar{R}_{1}, 3\bar{R}_{1}\}, \\ \hat{R}_{2}^{\tau} \triangleq & \operatorname{diag}\{\bar{R}_{2}, 3\varphi_{\tau}\bar{R}_{2}\}, \ \hat{R}_{22}^{\tau} \triangleq & \operatorname{diag}\{\bar{R}_{22}, 3\varphi_{\tau}\bar{R}_{22}\} \\ & (\varphi_{\tau} \triangleq (\tau + 1)/(\tau - 1) & (\tau > 1), \ \varphi_{1} \triangleq 1). \end{split}$$

Then, for any  $\phi(k) \in \mathscr{X}_{\rho}$  satisfying  $V_2(0) \leq 1$ , the system (1) can be exponentially stabilized by the controller (5) with  $K = YX^{-T}$ .

*Proof:* If the LMIs (40)-(42) are feasible, it can be seen that the matrix X is invertible. Then, we can define

$$\begin{cases}
P_{\alpha} \triangleq \check{X}^{-1}\bar{P}_{\alpha}\check{X}^{-T} \; (\check{X} \triangleq \operatorname{diag}\{X, X, X\}), \\
Q_{\alpha} \triangleq X^{-1}\bar{Q}_{\alpha}X^{-T}, \; S_{\alpha j} \triangleq X^{-1}\bar{S}_{\alpha j}X^{-T}, \\
R_{\alpha j} \triangleq X^{-1}\bar{R}_{\alpha j}X^{-T}, \; Z_{\alpha} \triangleq X^{-1}\bar{Z}_{\alpha}X^{-T}, \\
T_{i1} \triangleq X^{-1}, \; T_{i2} \triangleq \delta_{i}X^{-1}, \; \alpha, j = 1, 2, \; i = 1, 2, 3, \\
M_{1} \triangleq \hat{X}^{-1}\bar{M}_{1}\hat{X}^{-T} \; (\hat{X} \triangleq \operatorname{diag}\{X, X\}), \\
M_{2} \triangleq X^{-1}\bar{M}_{2}X^{-T}, \; H \triangleq \bar{H}^{-T}, \; K \triangleq YX^{-T}, \\
U \triangleq \bar{U}X^{-T}, \; V \triangleq \bar{V}X^{-T}, \; W \triangleq \bar{W}X^{-T}.
\end{cases}$$
(45)

Performing some congruence transformations to LMIs (39)-(44) (see [2], [28]), and using the notations in (45), the matrix inequalities (14)-(19) in Theorem 1 can be, respectively, obtained.

If the input delay  $\tau_k$  is not incorporated in the system (1), the corresponding closed-loop system can be written as

$$x(k+1) = \sum_{s=1}^{2^{m}} \varpi_{s}^{k} \left\{ [A + B(D_{s}K + D_{s}^{-}U)]x(k) + (A_{d} + BD_{s}^{-}V)\sum_{i=1}^{+\infty} \mu_{i}x(k-i) + BD_{s}^{-}W\sum_{j=1}^{+\infty} \mu_{j}\sum_{i=k-j}^{k-1} x(i) \right\}, \ k \ge 0.$$
(46)

Choose the following augmented Lyapunov functional:

$$\hat{V}(k) = \hat{\eta}^{T}(k) \operatorname{diag} \{X^{-1}, X^{-1}\} \bar{P} \operatorname{diag} \{X^{-T}, X^{-T}\} \hat{\eta}(k) + \sum_{j=1}^{+\infty} \mu_{j} \sum_{i=k-j}^{k-1} \lambda_{1}^{k-i-1} x^{T}(i) X^{-1} \bar{S}_{1} X^{-T} x(i) + \sum_{l=1}^{+\infty} \mu_{l} \sum_{j=1}^{l} \sum_{i=k-j}^{k-1} \lambda_{1}^{k-i-1} x^{T}(i) X^{-1} \bar{S}_{2} X^{-T} x(i) + \sum_{l=1}^{+\infty} \mu_{l} \sum_{j=1}^{l} \sum_{i=k-j}^{k-1} \lambda_{1}^{k-i-1} y^{T}(i) X^{-1} \bar{Z} X^{-T} y(i)$$
(47)

where  $\hat{\eta}(k) = \begin{bmatrix} x^T(k) & \sum_{j=1}^{+\infty} \mu_j \sum_{i=k-j}^{k-1} x^T(i) \end{bmatrix}^T$ . Then, it is easy to obtain the following control design condition.

Corollary 1: Let the scalars  $0 < \lambda_1 \leq 1$  and  $\delta \neq 0$  be given. Assume that there exist matrices  $0 < \bar{P} \in \mathbb{R}^{2n \times 2n}, 0 < \bar{S}_1 \in \mathbb{R}^{n \times n}, 0 < \bar{S}_2 \in \mathbb{R}^{n \times n}, 0 < \bar{Z} \in \mathbb{R}^{n \times n}, X \in \mathbb{R}^{n \times n}, Y \in \mathbb{R}^{m \times n}, \bar{U} \in \mathbb{R}^{\widehat{m} \times n}, \bar{V} \in \mathbb{R}^{\widehat{m} \times n}, \bar{W} \in \mathbb{R}^{\widehat{m} \times n}$ , such that for  $\forall s \in [1, 2^m], \forall l \in [1, \widehat{m}]$ , the following LMIs are feasible:

$$\hat{\Xi}_{1}(s) \triangleq \hat{\Gamma}_{1}^{T} \bar{P} \hat{\Gamma}_{1} - \lambda_{1} \hat{\Gamma}_{2}^{T} \bar{P} \hat{\Gamma}_{2} + \hat{\Psi}_{1} - \hat{\Phi}_{1}^{T} (\bar{Z} / \tilde{\sigma}_{1}) \hat{\Phi}_{1} + \operatorname{Sym}(\hat{T}_{1} \hat{\Sigma}_{1}) < 0, \qquad (48)$$

$$\hat{\Xi}_2(l) \triangleq \begin{bmatrix} 1 & \hat{N}_{(l)} \\ \hat{N}_{(l)}^T & \operatorname{diag}\{\bar{P}, 0\} + \hat{\Psi}_2/\lambda_1 \end{bmatrix} \ge 0$$
(49)

where  $\hat{N}_{(l)} = [\bar{U}_{(l)} \ \bar{W}_{(l)} \ \bar{V}_{(l)}]$  and

$$\hat{\Gamma}_{1} = \begin{bmatrix} I & 0 & 0 & I \\ \kappa I & -I & I & 0 \end{bmatrix}, \quad \hat{\Gamma}_{2} = \begin{bmatrix} I & 0 & 0 & 0 \\ 0 & 0 & I & 0 \end{bmatrix}, \\ \hat{\Phi}_{1} = \begin{bmatrix} \kappa I & -I & 0_{n \times 2n} \end{bmatrix}, \quad \hat{T}_{1} = \begin{bmatrix} I & 0_{n \times 2n} & \delta I \end{bmatrix}^{T}, \\ \hat{\Sigma}_{1} = \begin{bmatrix} (A - I)X^{T} + B(\mathcal{D}_{s}^{-}\bar{U} + D_{s}Y) \\ A_{d}X^{T} + B\mathcal{D}_{s}^{-}\bar{V} & B\mathcal{D}_{s}^{-}\bar{W} & -X^{T} \end{bmatrix}, \\ \hat{\Psi}_{1} = \operatorname{diag}\{\kappa\bar{S}_{1} + \sigma\bar{S}_{2}, -\bar{S}_{1}/\tilde{\kappa}_{1}, -\bar{S}_{2}/\tilde{\sigma}_{1}, \sigma\bar{Z}\}, \\ \hat{\Psi}_{2} = \operatorname{diag}\{0, \bar{S}_{2}/\tilde{\sigma}_{1}, \bar{S}_{1}/\tilde{\kappa}_{1}\} + (1/\tilde{\sigma}_{1}) \\ \times [\kappa I & 0 & -I \end{bmatrix}^{T}\bar{Z}[\kappa I & 0 & -I].$$

Then, for any  $\phi(k) \in \mathscr{X}_{\rho}$  satisfying  $\hat{V}_1(0) \leq 1$ , the system (1) can be exponentially stabilized by (5) with  $K = YX^{-T}$ .

*Remark 5:* In recent years, the local stabilization problem has been sufficiently addressed for discrete-time systems with time delays under saturating actuators [2], [16], [17], [19]. Different from such recent results where a single discrete state delay is involved, our obtained results are concerned with the discrete-time systems with *both* distributed state delay *and* fast-varying input delays. In fact, distributed delays are often encountered in various applications such as engineering systems, traffic flow models, biological systems and neural networks [4], [18], [22]. Moreover, almost all practical control

systems are subject to input delays (e.g., actuator and transmission delays). The results presented in Theorem 2 and Corollary 1 serve as indispensable complements of the existing results. For systems with finite distributed delay  $\sum_{i=1}^{h} \mu_i x(k-i)$ , the corresponding results can be readily obtained by revising the notations of  $\kappa$ ,  $\sigma$ ,  $\tilde{\kappa}_{\alpha}$  and  $\tilde{\sigma}_{\alpha}$ .

For the case that the distributed state delay is not involved in (1), one can select the following Lyapunov functional:

$$\check{V}(k) = \begin{cases} \check{V}_1(k), \ k \in [k^*, +\infty), \\ \check{V}_2(k), \ k \in [0, k^*) \end{cases}$$
(50)

where

$$\begin{split} \check{V}_{\alpha}(k) = \check{\eta}^{T}(k) \mathrm{diag}\{X^{-1}, X^{-1}\} \bar{P}_{\alpha} \mathrm{diag}\{X^{-T}, X^{-T}\} \check{\eta}(k) \\ + \sum_{i=k-\tau}^{k-1} \lambda_{\alpha}^{k-i-1} x^{T}(i) X^{-1} \bar{Q}_{\alpha} X^{-T} x(i) \\ + \tau \sum_{j=-\tau}^{-1} \sum_{i=k+j}^{k-1} \lambda_{\alpha}^{k-i-1} y^{T}(i) X^{-1} \bar{R}_{\alpha} X^{-T} y(i) \end{split}$$

with  $\check{\eta}(k) = \begin{bmatrix} x^T(k) & \sum_{i=k-\tau}^{k-1} x^T(i) \end{bmatrix}^T$  and  $\bar{R}_{\alpha} = \bar{R}_{\alpha 1} + \bar{R}_{\alpha 2}$ . Then, the following sufficient condition is readily obtained.

Corollary 2: Let the scalars  $0 < \lambda_1 \leq 1, \lambda_2 > 1, \nu > 0, \delta_i \neq 0$ (i = 1, 2, 3) and the integer  $k^* \geq 1$  be given. Assume that there exist matrices  $0 < \bar{P}_{\alpha} \in \mathbb{R}^{2n \times 2n}, 0 < \bar{Q}_{\alpha} \in \mathbb{R}^{n \times n}, 0 < \bar{R}_{\alpha j} \in \mathbb{R}^{n \times n}$  $(\alpha, j = 1, 2), \bar{M}_1 \in \mathbb{R}^{2n \times 2n}, \bar{M}_2 \in \mathbb{R}^{n \times n}, X \in \mathbb{R}^{n \times n}, Y \in \mathbb{R}^{m \times n}, \bar{U} \in \mathbb{R}^{\vec{m} \times n}$ , and the diagonal matrix  $0 < \bar{H} \in \mathbb{R}^{m \times m}$  such that, for  $\tau_k = 0, \tau, \forall s \in [1, 2^m], \forall l \in [1, \hat{m}]$ , the LMIs (39) and

$$\check{\Xi}_{1}(\tau_{k},s) \triangleq \check{\Gamma}_{1}^{T}\bar{P}_{1}\check{\Gamma}_{1} - \lambda_{1}\check{\Gamma}_{2}^{T}\bar{P}_{1}\check{\Gamma}_{2} + \check{\Psi}_{1} 
- \lambda_{1}^{T}\check{\Phi}_{1}^{T}\bar{\Lambda}_{1}\check{\Phi}_{1} + \operatorname{Sym}(\check{T}_{1}\check{\Sigma}_{1}) < 0, \qquad (51)$$

$$\check{\Xi}_{2} \triangleq \check{\Gamma}_{3}^{T}\bar{P}_{2}\check{\Gamma}_{3} - \lambda_{2}(\check{\Gamma}_{4}^{T}\bar{P}_{2}\check{\Gamma}_{4} + \check{\Phi}_{2}^{T}\bar{\Lambda}_{2}\check{\Phi}_{2}) + \check{\Psi}_{2}$$

$$\begin{aligned} z_2 &\equiv \Gamma_3^* P_2 \Gamma_3 - \lambda_2 (\Gamma_4^* P_2 \Gamma_4 + \Phi_2^* \Lambda_2 \Phi_2) + \Psi_2 \\ &- \lambda_2 \check{\Phi}_3^T \hat{R}_{22}^* \check{\Phi}_3 + \operatorname{Sym}(\check{T}_2 \check{\Sigma}_2 + \check{T}_4 \check{\Sigma}_4) < 0, \end{aligned} \tag{52}$$

$$\check{\Xi}_{3} \triangleq \check{\Gamma}_{5}^{T} \bar{P}_{2} \check{\Gamma}_{5} - \lambda_{2} \check{\Gamma}_{6}^{T} \bar{P}_{2} \check{\Gamma}_{6} + \check{\Psi}_{3} - \lambda_{2} \check{\Phi}_{4}^{T} \hat{R}_{5}^{T} \check{\Phi}_{4} + \operatorname{Sym}(\check{T}_{2} \check{\Sigma}_{2}) < 0$$
(53)

$$\bar{P}_1 < \nu \bar{P}_2, \ \bar{Q}_1 < \nu \bar{Q}_2, \ \bar{R}_{1i} < \nu \bar{R}_{2i}, \ i = 1, 2.$$
 (54)

$$\check{\Xi}_{4}(l) \triangleq \begin{bmatrix} 1/(\nu\lambda_{2}^{k^{*}}) & [\bar{U}_{(l)} \ 0] \\ [\bar{U}_{(l)} \ 0]^{T} & \bar{P}_{1} + \lambda_{1}^{\tau-1}\check{\Psi}_{4} \end{bmatrix} \ge 0$$
(55)

hold, where  $\hat{R}_1$ ,  $\hat{R}_2^{\tau}$ ,  $\hat{R}_{22}^{\tau}$  are denoted in Theorem 2 and

$$\begin{split} \check{\Gamma}_1 &= \begin{bmatrix} I & 0 & 0 & 0 & 0 & I \\ 0 & -I & -I & \check{\tau}_k I & \hat{\tau}_k I & 0 \end{bmatrix}, \\ \check{\Gamma}_2 &= \begin{bmatrix} I & 0 & 0 & 0 & 0 & 0 \\ -I & -I & 0 & \check{\tau}_k I & \hat{\tau}_k I & 0 \end{bmatrix}, \\ \check{\Gamma}_3 &= \begin{bmatrix} I & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & -I & \check{\tau} I & 0 & 0 \end{bmatrix}, \\ \check{\Gamma}_4 &= \begin{bmatrix} I & 0 & 0 & 0 & 0 & 0 \\ -I & 0 & 0 & \check{\tau} I & 0 & 0 \times (n+m) \\ 0 & -I & \check{\tau} I & 0 \end{bmatrix}, \\ \check{\Gamma}_5 &= \begin{bmatrix} I & 0 & 0 & I \\ 0 & -I & \check{\tau} I & 0 \end{bmatrix}, \\ \check{\Gamma}_6 &= \begin{bmatrix} I & 0 & 0 & 0 \\ -I & 0 & \check{\tau} I & 0 \end{bmatrix}, \\ \check{\Phi}_1 &= \begin{bmatrix} I & -I & 0 & 0 & 0 & 0 \\ I & I & 0 & -2I & 0 & 0 \\ 0 & I & -I & 0 & 0 & 0 \\ 0 & I & -I & 0 & 0 & 0 \\ 0 & I & -I & 0 & 0 \\ 0 & I & -I & 0 & 0 \\ 0 & I & -I & 0 & 0 \\ 0 & I & -I & 0 & 0 \\ 0 & I & -I & 0 & 0 \\ 0 & I & -I & 0 & 0 \\ 0 & 0 & 0 & 0 \\ \check{\Phi}_3 &= \begin{bmatrix} I & 0 & -I & 0 & 0 \\ I & 0 & I & -2I & 0 \\ I & 0 & I & -2I & 0 \\ 0 & 0 & 0 & 0 \\ \end{bmatrix}, \end{split}$$

$$\begin{split} \check{\Phi}_4 &= \begin{bmatrix} I & -I & 0 & 0 \\ I & I & -2I & 0 \end{bmatrix}, \ \check{\Phi}_5 &= \begin{bmatrix} \tau I & -I \end{bmatrix}, \\ \check{\Psi}_1 &= \text{diag}\{\bar{Q}_1, 0, -\lambda_1^{\tau}\bar{Q}_1, 0, 0, \tau^2\bar{R}_1\}, \\ \check{\Psi}_2 &= \text{diag}\{\bar{Q}_2, 0, -\lambda_2^{\tau}\bar{Q}_2, 0, \tau^2\bar{R}_2, 0\}, \\ \check{\Psi}_3 &= \text{diag}\{\bar{Q}_2, -\lambda_2^{\tau}\bar{Q}_2, 0, \tau^2\bar{R}_2\}, \\ \check{\Psi}_4 &= \text{diag}\{0, \bar{Q}_1/\tau\} + (2/\tilde{\tau})\check{\Phi}_5^T\bar{R}_1\check{\Phi}_5, \\ \bar{\Psi}_4 &= \text{diag}\{0, \lambda_1^{\tau}\bar{Q}_1/\tau, \bar{S}_{12}/\tilde{\sigma}_1, \bar{S}_{11}/\tilde{\kappa}_1\} \\ &+ (2\lambda_1^{\tau}/\tilde{\tau})\Phi_8^T\bar{R}_1\Phi_8 + (1/\tilde{\sigma}_1)\Phi_9\bar{Z}_1\Phi_9, \\ \check{T}_1 &= \begin{bmatrix} I & 0_{n\times 4n} & \delta_1I \end{bmatrix}^T, \ \check{T}_3 &= \begin{bmatrix} I & 0_{n\times 2n} & \delta_3I \end{bmatrix}^T, \\ \check{T}_2 &= \begin{bmatrix} I & 0_{n\times 3n} & \delta_2I & 0 \end{bmatrix}^T, \ \check{T}_4 &= \begin{bmatrix} 0_{m\times 5n} & I \end{bmatrix}^T, \\ \check{\Sigma}_1 &= \begin{bmatrix} (A-I)X^T & BD_s\bar{U} & BD_sY & 0_{n\times 3n} \\ &- X^T \end{bmatrix}, \ \check{\Sigma}_4 &= \begin{bmatrix} 0_{m\times n} & Y & 0_{m\times 3n} & -\bar{H} \end{bmatrix}, \\ \check{\Sigma}_2 &= \begin{bmatrix} (A-I)X^T & BY & 0_{n\times 2n} & -X^T & -B\bar{H} \end{bmatrix}, \\ \check{\Sigma}_3 &= \begin{bmatrix} (A-I)X^T & 0_{n\times 2n} & -X^T \end{bmatrix}. \end{split}$$

Then, for any  $\phi(k) \in \mathscr{X}_{\rho}$  satisfying  $\check{V}_2(0) \leq 1$ , the system (1) can be exponentially stabilized by (5) with  $K = YX^{-T}$ .

*Remark 6:* Recently, the local stabilization problem has been studied in [9] for linear input-delay systems with saturating actuators. Unlike [9], the constraint on the time-varying input delay is removed in this paper. Furthermore, the analysis approach proposed in [9] is no longer applicable because of the existence of multiple intervals with zero control signal and two dynamics within the interval  $[0, k^*)$ . In this paper, the more flexible piecewise Lyapunov functional (50) is proposed to characterize the possible state evolutions of different dynamics within the whole time interval. Moreover, different from the techniques used in [9], the augmented Lyapunov functional, the Wirtinger-based inequality and the current-state-dependent polytopic model are utilized together in this paper to reduce the conservatism.

Remark 7: In case the saturation level is non-unity, that is, sat $(u_l)$ = sgn $(u_l)$ min $\{|u_l|, \bar{u}_l\}$ , the matrices B and Y in Theorem 2 and Corollaries 1-2 should be substituted by  $\tilde{B} = [\bar{u}_1 b_1 \ \bar{u}_2 b_2 \cdots \bar{u}_m b_m]$ and  $\tilde{Y} = [y_1^T / \bar{u}_1 \ y_2^T / \bar{u}_2 \cdots y_m^T / \bar{u}_m]^T$ , respectively, where  $b_l$  is the l-th column of B and  $y_l$  is the l-th row of Y. Also, if the integer  $k^*$ is unknown,  $k^*$  in LMIs (44) and (55) should be replaced by  $\tau$ .

In the subsequent part, we will address the maximization problem of the initial condition set  $\mathscr{X}_{\rho}$ . First, let us introduce the LMI

$$\bar{P}_2 \leq \text{diag}\{\mathcal{J}_1, \mathcal{J}_2, \mathcal{J}_3\} \ (\mathcal{J}_i > 0, i = 1, 2, 3).$$
 (56)

Using (56) and Lemma 2, and noting (45), it follows that

$$V_{2}(0) \leq \left[\lambda_{M}(X^{-1}\mathcal{J}_{1}X^{-T}) + \tau^{2}\lambda_{M}(X^{-1}\mathcal{J}_{2}X^{-T}) + \sigma^{2}\lambda_{M}(X^{-1}\mathcal{J}_{3}X^{-T}) + \varphi_{1}\lambda_{M}(X^{-1}\bar{Q}_{2}X^{-T}) + \varphi_{22}\lambda_{M}(X^{-1}\bar{S}_{21}X^{-T}) + \varphi_{32}\lambda_{M}(X^{-1}\bar{S}_{22}X^{-T})\right]\rho_{1}^{2} + \left[\varphi_{4}\lambda_{M}(X^{-1}\bar{R}_{2}X^{-T}) + \varphi_{32}\lambda_{M}(X^{-1}\bar{Z}_{2}X^{-T})\right]\rho_{2}^{2}.$$
 (57)

To estimate  $\hat{V}(0)$  in Corollary 1, we introduce the LMI

$$P \le \operatorname{diag}\{\mathcal{J}_1, \mathcal{J}_2\} \ (\mathcal{J}_j > 0, j = 1, 2).$$
 (58)

Correspondingly,  $\hat{V}(0)$  can be enlarged as follows:

$$\hat{V}(0) \leq [\lambda_{M}(X^{-1}\mathcal{J}_{1}X^{-T}) + \sigma^{2}\lambda_{M}(X^{-1}\mathcal{J}_{2}X^{-T}) \\
+ \varphi_{21}\lambda_{M}(X^{-1}\bar{S}_{1}X^{-T}) + \varphi_{31}\lambda_{M}(X^{-1}\bar{S}_{2}X^{-T})] \\
\times \rho_{1}^{2} + \varphi_{31}\lambda_{M}(X^{-1}\bar{Z}X^{-T})\rho_{2}^{2}$$
(59)

Next, we introduce the matrix inequality  $X^{-1}X^{-T} \leq xI$  (x > 0 is a scalar) [1], which can be guaranteed by the following LMI:

$$\begin{bmatrix} xI & I\\ I & X + X^T - I \end{bmatrix} \ge 0.$$
 (60)

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Moreover, we introduce the following two sets of LMIs:

$$\begin{cases} \mathcal{J}_i \le p_i I, \ i = 1, 2, 3, \ \bar{Q}_2 \le q I, \\ \bar{S}_{0,i} \le s, I = \bar{R}_{0,i} \le r, I = i - 1, 2, \ \bar{Z}_0 \le r I \end{cases}$$
(61)

$$\begin{array}{l} (S_{2j} \leq S_j I, \ H_{2j} \leq I_j I, \ j = 1, 2, \ Z_2 \leq Z I, \\ \mathcal{J}_j \leq p_j I, \ \bar{S}_j \leq s_j I, \ i = 1, 2, \ \bar{Z} \leq Z I. \end{array}$$

Then, it is seen that the maximization of the admissible initial condition set  $\mathscr{X}_{\rho}$  in Theorem 2 and Corollary 1 can be, respectively, formulated by the following convex optimization problems:

$$P_{\alpha}, \bar{S}_{\alpha j}, \bar{Z}_{\alpha}, X, Y, \bar{U}, \bar{V}, \bar{W}, \mathcal{J}_{j}, x, p_{j}, s_{j}, z$$
LMIs in Corollary 1 and LMIs (58), (60), (62) hold

where  $\chi_1 = \epsilon x + p_1 + \tau^2 p_2 + \sigma^2 p_3 + \varphi_1 q + \varphi_{22} s_1 + \varphi_{32} s_2 + \varphi_4 r_1 + \varphi_4 r_2 + \varphi_{32} z$  and  $\chi_2 = \epsilon x + p_1 + \sigma^2 p_2 + \varphi_{21} s_1 + \varphi_{31} s_2 + \varphi_{31} z$ (the scalar  $\epsilon > 0$  is a weighting parameter).

*Remark 8:* In this paper, our obtained results are based on the novel polytopic models. If the same research is performed by using traditional polytopic model, the corresponding optimization problems are directly obtained by setting  $\bar{V} = \bar{W} = 0$  in (40) and (48), which are referred to as *Prob.1*' and *Prob.2*', respectively, in this paper.

For the case that the system (1) has a single input delay where u(k) = 0 for k < 0, we can see that the solution of (1) does not depend on the values of x(k) for  $k \le 0$  [9]. Then, we can define  $x(k) = \phi(k) = x_0, -\tau \le k \le 0$ . In this case, we assume that the initial condition  $x_0$  belongs to an ellipsoid as follows [9]:

$$\mathscr{E} \triangleq \{ x_0 \in \mathbb{R}^n : x_0^T P x_0 \le 1 \}.$$
(63)

From (50), it is seen that  $\check{V}_2(0)$  can be written as

$$\check{V}_2(0) = x_0^T X^{-1} \tilde{P} X^{-T} x_0 \triangleq x_0^T P x_0.$$
(64)

where  $\tilde{P} = \begin{bmatrix} I & \tau I \end{bmatrix} \bar{P}_2 \begin{bmatrix} I & \tau I \end{bmatrix}^T + \varphi_1 \bar{Q}_2$ . Let the LMI

$$\tilde{P} = \begin{bmatrix} I & \tau I \end{bmatrix} \bar{P}_2 \begin{bmatrix} I & \tau I \end{bmatrix}^T + \varphi_1 \bar{Q}_2 \le pI$$
(65)

be given. Then, the maximization of the admissible initial condition set  $\mathscr{E}$  involving in Corollary 2 can be described as follows:

Prob.3. 
$$\min_{\bar{P}_{\alpha}, \bar{Q}_{\alpha}, \bar{R}_{\alpha}, \bar{M}_{\alpha}, X, Y, \bar{U}, \bar{H}, x, p} \epsilon x + p, \ s.t.,$$
  
LMIs in Corollary 2 and LMIs (60), (65) hold.

## IV. NUMERICAL EXAMPLES

Example 1: Consider the system described by (1) where

$$A = \begin{bmatrix} 1.1 & 0.15 \\ 0.03 & 0.8 \end{bmatrix}, \ A_d = \begin{bmatrix} 0 & -0.1 \\ 0 & 0 \end{bmatrix}, \ B = \begin{bmatrix} 1 \\ 0.1 \end{bmatrix}$$
$$\mu^i = 2^{-i}, \ \bar{u}_1 = 15, \ 0 \le \tau_k = 2 + (-1)^k \le 3.$$

For this example, we first consider the special case without input delay. By solving Prob.2 and Prob.2' with  $\delta = 1.5$  and  $\epsilon = 2 \times 10^9$ , the admissible bounds of the initial condition set  $\mathscr{X}_{\rho}$  ( $\rho_1 = \rho_2 \triangleq \rho$ ) can be readily obtained for different  $\lambda_1 (> 1/2)$ , which are given in Table I. Recalling that Prob.2' is based on the traditional polytopic model, Table I shows that our proposed distributed-delay-dependent polytopic model is really effective in reducing the conservatism.

Next, we consider the more general case. By solving Prob.1 with  $\delta_1 = \delta_2 = \delta_3 = 4$ ,  $\epsilon = 2 \times 10^9$ ,  $\lambda_1 = 1$ ,  $\lambda_2 = 1.24$  and  $\nu = 0.97$ , it is found that the initial condition set  $\mathscr{X}_{\rho}$  can be bounded by  $2.0116\rho_1^2 + 1.7704\rho_2^2 \le 10^4$  with K = [-0.1695 - 0.0397]. In particular, we have  $\rho_1 \le 70.5065$  for the case  $\rho_2 = 0$ . Using the

TABLE IThe admissible bounds (ho) of the initial condition set  $\mathscr{X}_{
ho}$ 

| $\lambda_1$ | 0.90    | 0.93    | 0.95    | 0.87    | 1.0      |
|-------------|---------|---------|---------|---------|----------|
| Prob.2      | 74.3899 | 82.8034 | 89.5207 | 97.3417 | 110.5246 |
| Prob.2'     | 60.1129 | 69.6121 | 76.7917 | 84.7264 | 98.1399  |

TABLE II The admissible radii  $(\check{\rho})$  of the initial ball

| $\tau$ | 3      | 4      | 5      | 6      | 7          |
|--------|--------|--------|--------|--------|------------|
| [9]    | 2.2616 | 1.9100 | 1.5745 | 1.2022 | infeasible |
| Prob.3 | 2.5263 | 2.2977 | 2.0391 | 1.6340 | 0.9694     |

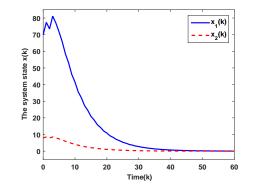


Fig. 1. The state evolutions of the closed-loop system.

above controller gain, the state evolution of this system is plotted in Fig. 1. In the simulation, the initial condition is selected as  $\phi(k) = \begin{bmatrix} 70 & 8 \end{bmatrix}^T \in \mathscr{X}_{\rho}$ . Noting that the open-loop system is not stable, it is seen from Fig. 1 that our proposed control scheme is effective.

Example 2: Consider the discrete time-delay system (1) where

$$A = \begin{bmatrix} 1.11 & -0.06\\ 0.05 & 0.9 \end{bmatrix}, B = \begin{bmatrix} 0.1\\ 0.1 \end{bmatrix}, A_d = 0,$$
  
$$\bar{u}_1 = 5, \ 0 \le \tau_k \le \tau, \ k^* \text{ is unknown.}$$

For this example, by solving the optimization problems in [9] and Prob.3 of this paper, we obtain some maximum admissible radii  $\check{\rho} \triangleq \sqrt{1/\lambda_M(P)}$  of the initial ball contained in the region of attraction, which are listed in Table II. In particular, when  $\tau = 5$ , we have

$$\mathcal{X} \triangleq \left\{ x_0 \in \mathbb{R}^2 : x_0^T \begin{bmatrix} 0.3683 & -0.1136 \\ -0.1136 & 0.0350 \end{bmatrix} x_0 \le 1 \right\} ([9]), \\ \mathscr{E} \triangleq \left\{ x_0 \in \mathbb{R}^2 : x_0^T \begin{bmatrix} 0.2196 & -0.0677 \\ -0.0677 & 0.0209 \end{bmatrix} x_0 \le 1 \right\} (\text{Prob.3}), \\ K = \left[ -2.1527 & 0.6683 \right] ([9]), \ K = \left[ -2.1379 & 0.6545 \right] (\text{Prob.3}).$$

In solving the optimization problem in [9], we select  $\lambda = 1$ ,  $\mu = 1.3$ ,  $\beta = 1$ ,  $\sigma = 0.001$  and  $\varepsilon = 12$ . In solving Prob.3, we choose  $\delta_1 = 10$ ,  $\delta_2 = \delta_3 = 8$ ,  $\lambda_1 = 1$ ,  $\lambda_2 = 1.23$ ,  $\nu = 0.9$  and  $\epsilon = 100$ . In Fig. 2, we plot part bounds of the sets  $\mathcal{X}$  and  $\mathscr{E}$  and some state trajectories of the closed-loop system. It is seen from Table II and Fig. 2 that our proposed result can provide a larger estimate of the region of attraction than that in [9]. Moreover, we notice that two trajectories starting not so far from the set  $\mathscr{E}$  diverge. In addition, it is worth mentioning that our result removes the constraint on  $\tau_k$ .

## V. CONCLUSIONS

In this paper, the local stabilization problem has been addressed for discrete-time systems with both distributed state delay and fast-

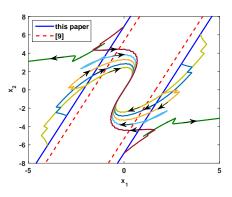


Fig. 2. Part bounds of the sets  $\mathcal{X}$  and  $\mathscr{E}$  and the state trajectories.

varying input delay under saturating actuators. By using a novel polytopic model, the piecewise Lyapunov functional and some summation inequalities, a local stabilization condition has been established in terms of LMIs. The special cases with either state delay or input delay have also been addressed. The proposed results in our paper can be easily extended to the case with discrete state delay. Also, it is interesting to consider the local stabilization problem for saturated systems with distributed input delay, which is our future work.

Here, it is worth mentioning that the techniques dealing with the time delays in this paper are somewhat conservative. By using the recent developed inequalities [11], [15], [26], [29], one can establish some more effective results, which is our further research. In addition, it is noted that the our results are based on the assumption that u(k) = 0 for k < 0. In case the past states can be used for feedback, the LMI  $\bar{\Lambda}_2 > 0$  in (39), LMIs (41)-(43), LMIs (52)-(54) should be removed, the term  $1/(\nu \lambda_2^{k^*})$  in LMIs (44) and (55) should be set as 1. Moreover, the constraints on  $\phi(k)$  in Theorem 2 and Corollary 2 should be revised as  $V_1(0) \leq 1$  and  $\tilde{V}_1(0) \leq 1$ , respectively.

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