Improving Controllability in RF-MEMS Switches using Resistive Damping

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ty in order to achieve soft Abstract. An efficient way to control the act velo landing and fewer bouncing phenomena is the tive nping. This control method e by Castaner and Senturia [1]. is also referred as charge drive and presented for fi Under charge control the Pull-in phe Constant Voltage controlled omenon of electrostatic actuators does not exist a irrent di s ideal, any position across the gap is stable. The main reason for t that the electrostatic force applied beha is always attractive and independent of of the actuator. Charge drive control incorporating const is mostly preferred to extend the travel ent so range of electrostatic mig s [2], [4], [5]. Nevertheless there are very few references in the lite ive control on RF MEMS. Recently ure abo charge hs for capacitive RF-MEMS, [6] and [7] published work based num present a learning algoi order to reduce fabrication variability using resistive se. Nevertheless none of them present any details on damping for ing and any results of such kind of applications. This how to im hent re da nts in detail t procedure in calculating the bias resistance of an RFwork p entire MEMS s controlled der resistive damping.

Key work charge dore control, RF-MÈMS switch, resistive damping, bouncing, contact force.

1. The Significance of Resistive Damping

The controllability of a switch is the key factor to reduce wear by minimizing the impact velocity. Despite the sophisticated design, adopting special cantilever shapes for contributed actuation force as well as utilizing fringing fields by making use of protruded electrodes, controllability still remains a difficult task which requires great thought and mathematical calculations. In case of a very stiff device, like the one which has been fabricated and presented by Guo, McGruer and Adams [8], the actuation control under resistive damping is the only way to achieve controllability. Due to the small switching time as well as the high actuation voltage, it is not practical to implement a tailored control pulse. Experimental results have shown that time intervals smaller than 1 μ s and pulses with slew-rate greater than 200V/ μ s are necessary in order to shape a tailored pulse for this switch, as the switching time is about 1.24 μ s when a sharp actuation pulse of 83V is applied. Even for the case that this fast and high in voltage pulse can be generated, there are other subjects like overshooting that they will possibly render problematic the control of the switch.

To eliminate bouncing phenomena, during the release phase of the switch, when the cantilever is oscillating within mechanical resonance frequency, the R_bC_{el} product must be equal to the period of the resonance frequency [9].

Very stiff devices [8], present high mechanical resonance frequency and make them appropriate for this kind of control as the time constant RC, which has been calculated for the pull-down phase, is near the period of the mechanical resonance frequency. Consequently, significant improvement in both switching operation phases of the switch is achieved. Thus, ontrol under this tive damping is the only practical solution for very fast RF-M MS switches where switching time and period of the resonance frequency are of the same order.

2. Applying Resistive Dampers to Imprese Controllability

The ohmic RF-MEMS switch of Fig. Tos been evaluated under the Coventorware software package examining controllability with and without resistive damping.



Fig. 1. The "NEU" ohmic RF MEMS Switch.

Initially, a transient analysis is performed under step pulse implementation with 83V amplitude, width $p_w=48\mu s$, rise time $t_r=1\mu s$ and fall time $t_f=1\mu s$. The amplitude of 83V has been calculated in order to be high enough to ensure immunity to switch parameters uncertainty due to the tolerances of the fabrication process, and low enough to ensure plenty of room for RF signal.

The switching time obtained under the above pulse conditions was some $1.7\mu s$ for the OFF-ON transition and around $1.4\mu s$ for the OFF-ON transition, as



shown in Fig. 2, the fastest ON and OFF switching time that can be achieved.

Fig. 2. Displacement under step pulse contrabode.

The same figure illustrates the bouncing prob ems durin the pull down (max. bounce=174µm) and release (max. bounce=25 ettling times are n) phases. Hig. observed also due to the stiffness of the can 000 N/m), which are some ever (k⁄ 11µs for the pull down phase and roughly the release-phase. In Fig. 3 other characteristics of the switch under step put implementation are illustrated, such as the contact area (11.566pn ce per contact area (2.53S condu which corresponds to a resistance o the contact force $(99.3\mu N)$. 0.39nd Control difficulties are illustrated also oncerns the high initial contact force (almost 496µN) due to the locity (around 65.9cm/sec). In order to ppact introduce resistive damp is necessary to be calculated. Having a bi resist g, calculated the capacital rode area (C_{el} =30fF) and with a pulse wit amplitude of 83V and rise of $t_r = 1 \mu s$, the bias resistance can be calculated has been extracted.



Fig. 3. Characteristics under step pulse control mode.

$$R_b C_{el} = t_r = 1 \mu m => R_b \approx 33 M\Omega \tag{1}$$

Figure 4 illustrates the characteristics of the switch under step pulse implementation with resistive damping. >



Fig. 4. Comparison between step pulse and restive Damping nodes.

The simulation results with $R_b=33M$ show excellent response of the switch during the pull down phase as elimina f the bouncing phenomena is observed as well as dramatic reduction of the initial impact force (the high impact velocity has been reduced to 13.2 from 6. cm/sec), with only a small increase in the switching time (3.47µ N). During the release phase a from ancing is observed too (174nm instead significant reduction of the applitude of of 255nm).

A comparison between stra-pulse and step pulse with resistive damping actuation modes is presented in 1g. 5.



Fig. 5. Characteristics under resistive damping control.

It is obvious that the control of the switch under resistive damping excels the corresponding with the step pulse in both OFF-ON and ON-OFF transitions

slightly sacrificing in the switching time. Finally, in Fig. 6, the power consumption of the switch under the previously mentioned actuation control modes is presented. It is clearly shown that under resistive control mode the switch requires much less power to be actuated.



Fig. 6. Power requirements under Pulsand resistive

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